Tools to Aid in Wiring Rule Generation for High Speed Interconnects

Paul Franzon

Slobodan Simovich

Michael Steer

Mark Basel

Sharad Mehrotra

Tom Mills

Picosecond Digital Electronics Laboratory
Department of Electrical and Computer Engineering
North Carolina State University
Box 7911
Raleigh, NC 27695

Abstract

This paper discusses the construction and application of two tools. The first tool, MetaSim, allows multiple simulation studies to be automatically specified, conducted, and analyzed in such a fashion as to produce wiring rules for layout tools. One of MetaSim's current engines, Transim is a convolution simulator that incorporates a robust, rapid, and accurate method for the simulation of coupled, lossy interconnect structures specified with frequency-dependent parameters.

1 Introduction

The high speed performance of most digital systems is now limited by distributed discontinuities and transmission line effects of packaging and interconnects, and not by the switching speed of semiconductor devices. Among the effects of these distributed circuit elements are: Delay and rise time degradation, ringing, crosstalk noise, metastates, and simultaneous switching noise. ¹

In many cases, the quest for higher performance is centering on choosing the most appropriate packaging technology, as well as layout considerations, not just increasing device clocking rates.

However, a number of difficulties are encountered as the circuit designer, layout engineer, and signal integrity engineer seek to design high speed systems.

One primary difficulty is the need for the signal integrity engineer to develop wiring rules to be used in the router [2]. These wiring rules constrain the lengths Precise wiring rules can not be generated from simple analytic expressions for propagation delay as no expression available today can accurately predict the effects of reflection and simultaneous switching noise on delay. Instead, the generation of wiring rules requires multiple simulation studies to be performed across an appropriate range of design and parametric variations. The simulation studies are analysed in order to characterise the transformation between the physical design space (P − lengths, spacings, etc.) and the electrical design space (E − delay, noise). This transformation is then inverted and applied to the placement in order to check it and to formulate wiring rules for the router. We will refer to this set of transformations as the "P↔E transformation".

Carrying out such studies traditionally requires a large amount of engineer time and can be error prone. MetaSim (Meta Simulator) automates the generation, conduct, and analysis of such simulation studies. In this paper, we describe the operation of MetaSim and give an example of its use.

MetaSim interfaces with three simulators, Spice, Casm, and Transim. Transim, solves the problem of

29th ACM/IEEE Design Automation Conference®

and spacings (and sometimes widths) on the board or multichip module so that delay and noise requirements are met. Care has to be taken in doing this. If the wiring rules are too lenient, many signal integrity problems will be detected during verification and the board will need extensive redesign. If the rules are too conservative, the router will be overconstrained and available routing area will be used inefficiently. Until recently rules of thumb often sufficed. However, as the rise times in complex CMOS boards and MCMs approach sub-1 ns ranges, the need for more precise wiring rules will dramatically increase.

¹This work was supported, in part, by the National Science Foundation under contract MIP-9017054

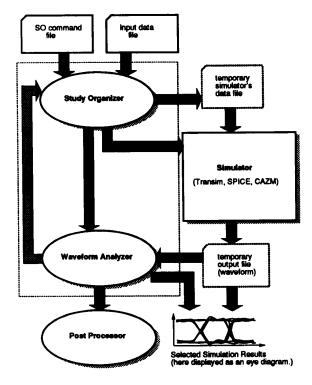


Figure 1: Organisation of MetaSim.

incorporating transient circuit simulation with nonlinear devices, with transmission line simulation in the frequency domain. For example, this capability is a requirement for thin film MCM interconnect once the skin effect starts becoming important (ie. when rise times start approaching around 200 ps [4].)

2 MetaSim: A Simulation Generator and Analyzer

MetaSim (figure 1) automates the <simulation set up → simulation execution → simulation results analysis> loop. It thus allows complex relationships between delay and noise requirements on one hand, and technology selection and physical design issues, on the other hand, to be explored with a minimum of human time. MetaSim consists of the following elements (Figure 1):

A study organiser, which allows variational studies requiring multiple simulation runs to be concisely specified and generated. It also includes specialised simulation file generators for Casm, SPICE and Transim.

- A waveform analyser that carries out signal integrity analysis on each set of simulation results and extracts delay and noise data as required.
- Post-processing software that enables relationships to be readily extracted from the analysed results, and selected waveforms to be retained and viewed.

The nature of these tools will be described further, in section 3, primarily by the way of an example.

In the following sections, we will first describe related work conducted elsewhere, and present an example of MetaSim's use.

2.1 Related Work

The existing approach to wiring rule generation is to make initial estimates and refine on them through multiple iterations of the \langle specify rules \rightarrow carry out route \rightarrow verify \rightarrow analyse simulations \rangle loop. Although some commercial tools automate part of the analysis they are not able to suggest how to improve the rules. Attempts at this, such as TLTS [7] make very simple suggestions such as 'shorten net' and 'remove receivers'. As better suggestions require that a characterisation of the $P \leftrightarrow E$ transformation exist anyway, it makes sense to use this transformation to form good rules initially for as many nets as is reasonable. This will greatly reduce the number of design iterations required, hopefully to one.

The closest existing tool to MetaSim is Adline, which was developed at NTT [5]. It is only available in Japan. Adline includes in its simulator, the ability to specify multiple simulations for some structures. With Adline, as with MetaSim, it is also possible to study the effect of process variations by automatically carrying out multiple simulations. The main distinction between Adline and MetaSim is that MetaSim offers far greater flexibility in the range of studies that can be undertaken. MetaSim can also be easily integrated with any simulator while Adline is integral with one simulator.

3 Example of the Application of MetaSim

MetaSim is applied before layout by taking a net, or set of nets, that are likely to require layout constraints and conducting studies on them. For example, here, we will consider the set of nets that have one output and four inputs, and the output may, or may not be,

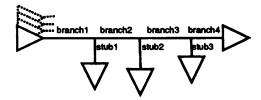


Figure 2: Network topology to be analysed. The output driving the net might have up to three other outputs switching with it. The user can specify any circuit model for the ground and power connections and can specify driver skew.

part of four simultaneous switching outputs. The interconnection network topology is described as a set of branches and stubs, in a daisy chain configuration, as shown in figure 2. The circuit technology is $0.8\mu m$ CMOS, and the packaging technology, a polyimide-dielectric thin film MCM, with copper interconnect, and non-fan-out-TAB chip attach.

For the purposes of this paper all of the branches shown in figure 2 will be set at the same length as will the stubs. This is only done to make the results presentable. Part of the study command file that specified the studies that produced the results presented here is shown in figure 3. Part of the 'data file' is also presented. In this study, three network related components are varied independently, the driver circuit model, the branch length and stub length, in order to specify a total of 32 simulation runs. The command language is actually a formally defined programming language with many constructs other than the ones shown including random generation constructs.

The 'data file' shown in figure 1 is a generic simulation specification file that lists all of the elements to be simulated as a network of objects (eg. a branch in a transmission line is an object) with certain properties associated with them. The "\$" variables that appear in the command file also appear in the data file. The study organiser then produces appropriate substitutions for each simulation run and passes the modified data file to the program that generates the actual simulation file (or directly to Transim, as a Transim file is already described in object-orientated terms). Discontinuities, such as vias, bends, and crossovers can also be specified in the data file, as can coupled lines and connectors.

The 'do' statement specifies what analysis needs to be carried out on the simulation results. In this case we wish to know the peak undershoot at node "5" (the last receiver in the chain) and the delay from the driver output, node "1" to this receiver. We are Extract from command file:

```
contr
  for ($branch = 1e-2 4e-2 1e-2 incr)
  for ($stub= 1e-3 2.9e-2 4e-3 incr)
  do{
    undsht = "5"
    delay = "1" "5" 0.23
    delay = "1" "5" 0.734
  }
```

Extract from 'data' (generic simulation specification)

Branch 1:

Brch_type_1 bstrip Brch_length_1 \$branch Height1&_1 10e-6 Height1B_1 20e-6 Thick1_1 5e-6 Width1_1 17e-6 Er1_1 3.5 Er2_1 3.5 Brch_load_1 non-loaded

Figure 3: Part of the command and data files used to specify the study described here.

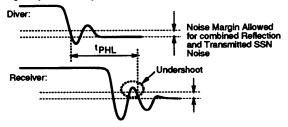


Figure 4: Definition of delay to include settling time for noise.

not interested in the 50% delay, however. As this is a data line, we wish to know when it is safe to latch the data, that is when the noise on the signal line has settled to the point that the combined reflection and transmitted simultaneous switching noise has settled to less than that fraction of the noise margin that the noise budget allows for it, as shown in figure 4. In the command file shown in figure 3, the maximum allowed combined reflection and transmitted simultaneous switching noise for a high-to-low transition is 23% of the voltage swing. Many other analyses can be done, including overshoot and metastate analysis. It is also possible to specify other noise budget and delay specifications.

Figure 5 shows some of the results obtained from running MetaSim with Casm as the simulation engine. An important result demonstrated through these examples is that despite the non-regular shapes generally

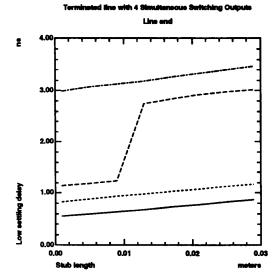


Figure 5: Total high-to-low delay between the ends of the line for different branch and stub lengths, for series terminated lines. (Solid line = 1cm branch lengths, short dashes = 2cm, long dashes = 3 cm, dash-dot = 4 cm.)

found in simulation waveforms, the trends in the settling time plots are generally regular. The steps in the plots are the only irregularities. The steps arise when a change in length suddenly makes the peak undershoot voltage edge past the noise threshold.

This result is important because it means that, as long as these steps are located, it is not necessary to carry out a large number of simulations in order to characterise the physical design space being considered. It also means that piecewise linear equations could be fitted. All of the studies we have carried out so far verify this result (sometimes piecewise polynomial equations would be required, however) though we can not claim to have analysed every possible situation. Sensitivities would be expected to increase with pulse edge rate, for example.

The physical design tools would then use these results to adjust the placement and specify wiring rules. For example, if the total high-to-low delay between two chips at opposite ends of a net had to be less than 1.0 ns, then these results can be used by the placement tool to indicate a requirement for a parallel termination in order to control the excessive noise. If the two pins in question were just under 8 cm apart then a maximum stub length of 1 cm is permitted as long as the first to last pin length is 10 cm. The placement

tool can then determine if it is likely that the router can meet these constraints. If there is a high likelihood that it can, then these constraints are passed to the router as a wiring rule. If it is likely that it can not then the placement tool suggests an improved placement. Automating this procedure should be straightforward.

4 Transim

Transim uses a convolution technique to allow mixed simulation of linear elements specified in the frequency domain and non-linear elements that need to be simulated in the time domain. In this section of the paper, we will first briefly discuss the operation of the convolution simulation core before presenting its operation as a complete simulator.

4.1 Related Work

Several approaches to this problem have been tried with varying degrees of success. The Laplace Transform technique is good at handling multiconductor transmission line systems and while it is more efficient than the convolution technique, its limited ability to handle frequency dependent effects and problems dealing with nonlinear terminations restrict its' usefulness. The Method of Characteristics [1] is relatively easy to use with conventional circuit simulators, but it is restricted to lines described by RLGC parameters. Other techniques for simulating distributed element systems have recently been reviewed by Djordjevic et al [3] and Schutt-Aine and Mittra [6].

In one of the convolution techniques the frequency domain network description is converted to a timedomain description using a Fourier transform. While this technique can handle arbitrary lossy coupled networks, aliasing in the frequency-domain to timedomain transformation can cause appreciable errors in the simulated transient response.

4.2 Convolution Simulation Core

Our approach is similar in some respects to that of Djordjevic et al, expect that where Djordjevic et al use admittance (Y) parameters to characterise the frequency dependent network, we use the scattering (S) parameters. As the S parameters vary only between -1 and +1, we overcome the dynamic range problems with the other convolution method's approach. S parameters of a transmission line system

describe the relative amplitude and phase of the forward and backward traveling waves at each port and at each frequency on a transmission line of characteristic impedance Z_{m} .

Integration into a transient circuit simulator requires that the description of the transmission line system be in terms of total voltages rather than the traveling wave components implied by use of S parameters. This can be achieved by terminating each port in its reference impedance so ensuring that there are no reflections at the ports of the distributed network. Now the total transient response of any linear network to voltage sources at the ports of the network can be determined by convolving in time these voltages with the Dirac impulse voltage responses at each external port of the network. More explicitly, the total transient response at port i of a transmission line system to a total voltage $E_j(t)$ with output impedance Z_m at each port is [8]

$$V_{i}(t) = \sum_{j=1}^{N} \int_{-\infty}^{t} g_{ij}(t-\tau)E_{j}(\tau)d\tau$$
$$= \sum_{j=1}^{N} g_{ij}(t) \star E_{j}$$
(1)

Here the symbol \star denotes convolution, N is the number of external ports in the transmission line system and

$$g_{ij}(t) = \frac{1}{\Delta t} \mathbf{F}^{-1} [G_{ij}(\omega)] = V_i^-(t) / E_j(0)$$
 (2)

is the impulse response at port i and at time t to a Dirac delta source at port j and time sero, In (2)

$$G_{ij}(\omega) = \begin{cases} (1 + S_{ij}(\omega))/2 & i = j \\ S_{ij}(\omega)/2 & i \neq j \end{cases}$$
 (3)

is the frequency-domain Green's function from port j to port i at radian frequency ω and $\mathbf{F}^{-1}[\]$ is the inverse Fourier transform. Finally, $\Delta t = 0.5/f_{mas}$ is the time increment of $g_{ij}(t)$ where f_{mas} is the highest frequency component of $G_{ij}(\omega)$.

Based on these techniques, Transim has been implemented [8]. Essentially, the interconnect structures to be simulated, are partitioned into a collection of linear, frequency dependent parameter structures, and non-linear circuit structures (including time dependent voltage and driver sources). Simulation proceeds by minimising the difference between the estimated voltages at the terminals of the non-linear elements and the voltages at the terminals of the linear elements, as determined via convolution.

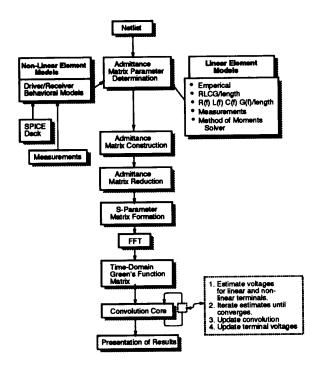


Figure 6: Complete Structure of Transim Simulator.

4.3 Operation as a Packaging Simulator

The complete structure of Transim is shown in Figure 6. Transim is a model-independent simulator. Instead modeled elements are represented internally as admittance matrices. Thus new models for both interconnect structures and drivers and receivers can be added quickly. From the Y-matrices corresponding to all of the modeled elements a complete Y-parameter matrix is constructed. An important step is the reduction of this matrix to a $M \times M$ admittance matrix, where M is the number of nonlinear nodes. After reduction, the Y-matrices are converted to S-matrices and then an FFT is carried out in order to obtain the Green's functions. These are then used in the convolution.

5 Conclusions

MetaSim can be used to characterise the transformation between the physical design space (length, spacing, etc.) and electrical design space (delay, noise) before any layout is done. The layout tools can then apply this transformation to the placement in order to suggest placement improvements and also to formu-

late wiring rules (routing constraints) for use by the router. MetaSim could also be used for other physical design scenarios where analytic delay and noise models are inadequate, such as high speed DRAM design, and packaging technology selection. It is very flexible and could be integrated with any simulator.

As it is used for a-priori analysis, MetaSim would conduct more simulations than a post-layout simulator would be required to. This suggests that increased run time might be an issue of concern. There are three counterpoints, however. First, in high speed designs, avoiding the need for multiple iterations of the design will save more user time and computer time than would be spent in using MetaSim up front, and will also lead to more efficient use of board space. Second, MetaSim is very efficient with the human designer's time. Third, initial results indicate that the P↔E transformation is generally regular and thus a good characterisation can be obtained with a small number of randomly generated results, as long as the irregular points are found.

Transim is one of the three simulation engines currently in MetaSim. It is a convolution-based simulation that has solved a number of the problems usually found in such simulators. It has been developed as a complete packaging simulator.

Acknowledgements

The authors would like to thank Jack Gregenkemper of Tandem, George Katopis and Jay Diepenbrock, both of IBM, and Steve Lipa of NCSU for discussions about circuit signal integrity requirements. Thanks also to Zaki Rakib, of Cadence, and Wayne Dai, of the University of California (Santa Crus), for discussions about CAE integration and physical design issues. We also gratefully acknowledge MCNC, particularly Wayne Detloff, for providing us with their $0.8\mu m$ CMOS driver and receiver simulation files.

References

- F.Y. Chang. The generalised method of characteristics for waveform relaxation analysis of lossy coupled transmission lines. In *IEEE MTT-S Di*gest, pages pp.821-826, 1989.
- [2] E.E. Davidson and G.A. Katopis. Package electrical design. In R.R. Tummala and E.J. Rymassewski, editors, Microlectronics Semiconductor Handbook, chapter 3. Van Nostrand Reinhold, 1989.

- [3] A.R. Dhordjevic, T.K. Sarkar, and R.F. Harrington. Time-domain response of multiconductor transmission lines. *Proc. IEEE*, 75:pp.743-764, June 1987.
- [4] B.K. Gilbert and W.L. Walters. Design options for digital multichip modules operating at high system clock rates. In *Proc. 1st ISHM MCM Conference*, 1992.
- [5] T. Mikasuki and N. Matsui. Statistical design technques for high-speed circuit boards. In IEEE/CHMT'90 IEMT Symposium, number IEEE no. CH2864-7/90/-0000-0185, pages 185–191, 1990.
- [6] J.E. Schutt-Aine and R. Mittra. Nonlinear transient analysis of coupled transmission lines. *IEEE Trans. Circuits and Systems*, Vol. CAS-36:pp. 959-967, July 1989.
- [7] E. Simoudis. A knowledge-based system for the evaluation and redesign of digital circuit networks. IEEE Trans on CAD, 8(3):pp.302-315, March 1989.
- [8] D. Winkelstein, M.B. Steer, and R. Pomerleau. Simulation of the transient response of arbitrary transmission line networks with nonlinear terminations. *IEEE Trans. Circuits and Systems*, pages pp.418-422, April 1991.