## Tools to Aid in Wiring Rule Generation for High Speed Interconnects NCSU-VLSI - 93- 08 Published in the 1992 Design Automation Conference

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#### Abstract

tegrity engineer to develop wiring rules to be used in This paper discusses the construction and application the router [2]. These wiring rules constrain the lengths tion of two tools. The first tool, MetaSim, allows mul-and spacings (and sometimes widths) on the board tiple simulation studies to be automatically specified; multichip module so that delay and noise requireconducted, and analyzed in such a fashion as to pro-ments are met. Care has to be taken in doing this. If duce wiring rules for layout tools. One of MetaSim's the wiring rules are too lenient, many signal integrity current engines, Transimis a convolution simulator problems will be detected during verification and the that incorporates a robust, rapid, and accurate method board will need extensive redesign. If the rules are tures specified with frequency-dependent parameters.

#### Introduction

for the simulation of coupled, lossy interconnect struteo conservative, the router will be overconstrained and available routing area will be used inefficiently. Until recently rules of thumb of ten suffeed. However, as the rise times in complex CMOS boards and MCMs approach sub-1 ns ranges, the need for more precise wiring rules will dramatically increase.

One primary difficulty is the need for the signal in-

The high speed performance of most digital sys-Precise wiring rules can not be generated fromsimtems is now limited by distributed discontinuitieplanchalytic expressions for propagation del ay as no extransmission line effects of packaging and interpression available today can accurately predict the efnects, and not by the switching speed of semicondufects of reflection and simultaneous switching noise on tor devices. Among the effects of these distribudeday. Instead, the generation of wiring rules requires circuit elements are: Delay and rise time degradatindit, i ple simulation studies to be performed across an ringing, crosstalk noise, metastates, and simultaampepuruspriate range of design and parametric variations. The si mul ati on studies are analyzed in order to charac-

In many cases, the quest for higher performance terize the transformation between the physical centering on choosing the most appropriate packagismpace (P-lengths, spacings, etc.) and the electrical technology, as well as layout considerations, notdgasgn space (E - delay, noise). This transformation increasing device clocking rates. is then inverted and applied to the placement in or-

However, a number of difficulties are encounterder to check it and to formulate wiring rules for the as the circuit designer, layout engineer, and signablution. We will refer to this set of transformations as tegrity engineer seek to design high speed systemshe "P↔E transformation".

<sup>&</sup>lt;sup>1</sup>This work was supported, in part, by the National Scie Carrying out such studies traditionally requires a Foundation under contract MIP-9017054 large amount of engineer time and can be error prone.

Figure 1: Organization of MetaSim

route  $\rightarrow$  verify  $\rightarrow$  analyze simulations > loop. though some commercial tools automate part of the

Japan. Adline includes in its simulator, the ability to specify multiple simulations for some structures. Wth Adline, as with MetaSim, it is also possible to study

the effect of process variations by automatically carrying out multiple simulations. The main distinction

For the purposes of this paper all of the branches

between Adline and MetaSimis that MetaSim offers

MetaSim (Meta Simulator) automates the generation analysis they are not able to suggest how to improve conduct, and analysis of such simulation studiestherules. Attempts at this, such as TLTS [7] make this paper, we describe the operation of MetaSimamedry simple suggestions such as 'shorten net' and 'regive an example of its use. move receivers'. As better suggestions require that

MetaSiminterfaces with three simulators. Spicecharacterization of the P↔E transformation exist Cazm, and Transim Transim, solves the problem of anyway, it makes sense to use this transformation to incorporating transient circuit simulation with nbooting good rules initially for as many nets as is reasonear devices, with transmission line simulation is ablieve. This will greatly reduce the number of design frequency domain. For example, this capability jistearations required, hopefully to one. requirement for thin film MCM interconnect once the The closest existing tool to MetaSimis Adline,

skin effect starts becoming important (ie. when rightsiech was developed at NTT [5]. It is only available in times start approaching around 200 ps [?].)

### MetaSim: A Simulation Generator and Analyzer

MetaSim (figure 1) automates the <simulation far greater flexibility in the range of studies that can set up → simulation execution → simulation resupet undertaken. MetaSimcan also be easily integrated analysis > loop. It thus allows complex relations in the annual simulator while Adline is integral with one between delay and noise requirements on one handsi mulator. and technology selection and physical design issues, on the other hand, to be explored with a minimum of human time. MetaSimconsists of the following el3- Example of the Application of ments (Figure 1): MataSi m

- Astudy organizer, which allows variational studies requiring multiple simulation runs to be coMetaSimis applied before layout by taking a net, or cisely specified and generated. It also inclarde of nets, that are likely to require layout constraints specialized simulation file generators for Cana, conducting studies on them For example, here, we will consider the set of nets that have one output SPICE and Transim
- and four inputs, and the output may, or may not be, A waveformanalyzer that carries out signal part of four simultaneous switching outputs. The integrity analysis on each set of simulation results nnection network topology is described as a set and extracts del ay and noise data as required of branches and stubs, in a daisy chain configuration,
- Post-processing software that enables relation, as shown in figure 2. The circuit technology is 0.8 µm ships to be readily extracted from the analyzed and the packaging technology, a polyimideresults, and selected waveforms to be retained and non-fer and TAP 1. viewed.

ample of MetaSim's use.

The nature of these tools will be described furthesthown in figure 2 will be set at the same length as will the stubs. This is only done to make the results section 3, primarily by the way of an example. In the following sections, we will first descrippeeseentable. Part of the study command file that speclated work conducted elsewhere, and present an eixfeed the studies that produced the results presented here is shown in figure 3. Part of the 'data file' is also presented. In this study, three network related com-Related Work ponents are varied independently, the driver circuit model, the branch length and stublength, in order to

The existing approach to wiring rule generation pescify a total of 32 simulation runs. The command to make initial estimates and refine on them throughanguage is actually a formally defined programming multiple iterations of the <specify rules  $\rightarrow$  carry anogutage with many constructs other than the ones

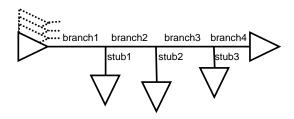


Figure 2: Network topology to be analyzed. The output dri ving the net might have up to three other outputs switching with it. The user can specify any circuit model for the ground and power connections and can specify driver skew.

Extract from command file:

```
contr
 for ($branch = 1e-2 \ 4e-2 \ 1e-2 \ incr)
 for ($stub= 1e-3 2.9e-2 4e-3 incr)
  do{
   undsht = "5"
    delay = "1" "5" 0.23
    delay = "1" "5" 0.734
```

fil e:

#### Branch 1:

Brch\_type\_1 bstrip Brch\_length\_1 \$branch Height1A\_1 10e-6 Height1B\_1 20e-6 Thick1\_1 5e-6 Width1\_1 17e-6 Er1\_1 3.5 Er2\_1 3.5 Brch\_load\_1 non-loaded

specify the study described here.

shown including randomgeneration constructs.

lation specification file that lists all of the elemplosts starre the only irregularities. The steps arise when be simulated as a network of objects (eg. a branchian change in length suddenly makes the peak undertransmission line is an object) with certain propehbioesvoltage edge past the noise threshold. associated with them The "\$" variables that appear This result is important because it means that, as in the command file also appear in the data file. Theong as these steps are located, it is not necessary to study organizer then produces appropriate substictauty out a large number of simulations in order to tions for each simulation run and passes the modified aracterize the physical design space being considdata file to the program that generates the actual simed. It also means that piecewise linear equations ulation file (or directly to Transim, as a Transim folded d be fitted. All of the studies we have carried out is already described in object-orientated terms) solf ar verify this result (sometimes piecewise polynocontinuities, such as vias, bends, and crossoversial apequations would be required, however) though we also be specified in the data file, as can coupled leames not claim to have analyzed every possible situaand connectors. tion. Sensitivities would be expected to increase with

The 'do' statement specifies what analysis neepdslse edge rate, for example. to be carried out on the simulation results. In thinkse physical design tools would then use these re-

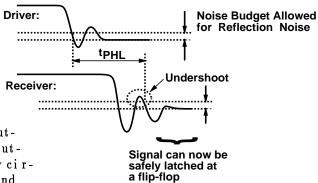
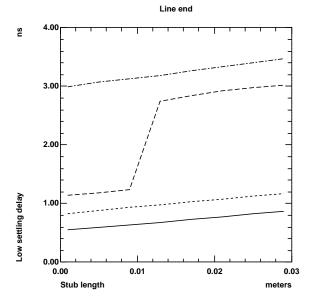


Figure 4: Definition of delay to include settling time for noise.

case we wish to know the peak undershoot at node "5" (the last receiver in the chain) and the delay from the driver output, node "1" to this receiver. We are not interested in the 50% delay, however. As this is a data line, we wish to know when it is safe to latch the data, that is when the noise on the signal line has settled to the point that the combined reflection and transmitted simultaneous switching noise has set-Extract from 'data' (generic simulation specificatient) to less than that fraction of the noise margin that the noise budget allows for it, as shown in figure 4. In the command file shown in figure 3, the maximum allowed combined reflection and transmitted simultaneous switching noise for a high-to-low transition is 23% of the voltage swing. Many other analyses can be done, including overshoot and metastate analysis. It is also possible to specify other noise budget and del ay speci fications.

Figure 5 shows some of the results obtained from Figure 3: Part of the command and data files used to unning MetaSimwith Cazmas the simulation engine. An important result demonstrated through these examples is that despite the non-regular shapes generally found in simulation waveforms, the trends in the set-The 'data file' shown in figure 1 is a generic simpling time plots are generally regular. The steps in the





terminated lines. (Solid line = 1cm brancihnlengthungthund ated transient response. short dashes = 2 cm, long dashes = 3 cm, dash-dot =4 cm.)

operation as a complete simulator.

#### 4.1 Related Work

Several approaches to this problem have been tri with varying degrees of success. The Laplace Tra form technique is good at handling multiconduct transmission line systems and while it is more cient than the convolution technique, its limited ity to handle frequency dependent effects and pro lems dealing with nonlinear terminations restriusefulness. The Method of Characteristics [1] is tively easy to use with conventional circuit simul but it is restricted to lines described by RLGC pa eters. Other techniques for simulating distribut ement systems have recently been reviewed by Djor jevic et al [3] and Schutt-Aine and Mittra [6].

In one of the convolution techniques the freque domain network description is converted to a time domain description using a Fourier transform. Wh this technique can handle arbitrary lossy coup

Figure 5: Total high-to-low delay between the two fks, aliasing in the frequency-domain to t the line for different branch and stublengt hao mat ion can cause appreciable err

#### 42 Convolution Simulation Core

sults to adjust the placement and specify wiring rules. For example, if the total high-to-low delay Oncert wepop proach is similar in some respects to the two chips at opposite ends of a net had to be of sDjohdjnevic et al, expect that where Djordjevi 1.0 ns, then these results can be used by the appl accempedati ttance (Y) parameters to characterize tool to indicate a requirement for a parallferle of a smit grade pendent network, we use the scatteri tion in order to control the excessive nois (S) I featrherms two rs. As the S parameters vary only be pins in question were just under 8 cm apart whehenn-ay and +1, we overcome the dynamic range the router as a wiring rule. If it is likely that it is man the the confice Z

# Transi m

forward.

maximum stublength of 1 cm is permitted as bentslæsms with the other convolution method's a the first to last pin length is 10 cm. The phasementS parameters of a transmission line sys tool can then determine if it is likely thadet begirbe ut the relative amplitude and phase of the can meet these constraints. If there is a waight alide to bake tward traveling waves at each port and hood that it can, then these constraints aratpeash drequency on a transmission line of charac then the placement tool suggests an improved Indtagration into a transient circuit simulate ment. Automating this procedure should be squissies hat the description of the transmission system be in terms of total voltages rather than

in its reference impedance so ensuring that ther no reflections at the ports of the distributed net Transim uses a convolution technique tNow thewtotal transient response of anylinear net mixed simulation of linear elements specified wilntake sources at the ports of the network car frequency domain and non-linear elements tdheattenmienled by convolving in time these voltages w to be simulated in the time domain. In this shetDiomao fimpulse voltage responses at each exte the paper, we will first briefly discuss the opposon tato if on those f network. More explicitly, the total the convolution simulation core before pressieemating siptos as at port i of a transmission line sy

traveling wave components implied by use of S para: eters. This can be achieved by terminating each p Figure 6: Complete Structure of Transim Si mulator functions. These are then used in the con to a total volt; a(g): What hout put impeda, nate Z lution. each port is [8]

$$egin{align} V_i(\ t) &= \sum_{m{j}\,=\,1}^N \int_{-\,\infty}^t g_i\ \{\!\!\{\ t- au\}\ E\!\!\!/\ t\ \}\ d au \ &= \sum_{m{j}\,=\,1}^N g_i\ \{\!\!\{\ t\ \}\ \star E\!\!\!/\ t\ \} \end{array}$$

Here the symbol  $\star$  denotes convolution, N is

$$g_{i} 
otin t) = \frac{1}{\Delta t} \mathbf{F}^{-1} [G_{i}(\omega)] = V(t) / E(0)$$
 (2)

is the impulse response at port i and at ti

$$G_{i} \ _{i} \omega) \ = \left\{ egin{array}{ll} \left( \ 1 + S \ _{i} \ \omega 
ight) \ /2 & i = j \ S_{i} \ _{i} \ \omega 
ight) \ /2 & i 
eq j \end{array} 
ight. \ \left( \ 3 \ 
ight)$$

the time increment (o) where of is the highest would be spent in using Meta Simup front, and wil frequency component (ob)G

dent voltage and driver sources). Simulation proceeds found. by minimizing the difference between the est Transidmis one of the three simulation engines of ments, as determined via convolution.

#### Operation as a Packaging Simulator

#### Acknowledgements

The complete structure of Transimis shown in Figure 6. Transimis a model-independent simula 17thue. alumihors would like to thank Jack Gregenkemstead modeled elements are represented intpeerrnafl Tyanadsem, George Katopis and Jay Diepenbrock admittance matrices. Thus new models for blooth hibfi-IBM, and Steve Lipa of NCSU for discussion terconnect structures and drivers and receia/boomstocain bait signal integrity requirements. T added quickly. From the Y-matrices correspandiontgot Zaki Rakib, of Cadence, and Wayne Dai, of all of the modeled elements a complete Y-pahanWentiewersity of California (Santa Cruz), for di matrix is constructed. An important step is other arborut CAE integration and physical design duction of this matrix to a M imes M admittanscreema-Wéalsogratefully acknowledge MCNC, partic trix, where M is the number of nonlinear nodeasr.l yAfWay me Detloff, for providing us with their 0. reduction, the Y-matrices are converted to CSMOMathrices and receiver simulation files.

#### Concl usi ons

MetaSim can be used to characterize the trans formation between the physical design space (len spacing, etc.) and electrical design space (delay before any layout is done. The layout tools can t apply this transformation to the placement in orde the greath placement improvements and also to form ber of external ports in the transmission hinge swiftig rules (routing constraints) for use router. MetaSimcould also be used for other physi design scenarios where analytic delay and noise r els are inadequate, such as high speed DRAMdesig: and packaging technology selection. It is very fle Phd to ut d be integrated with any simulator.

Dirac delta source at port j and time zero, In As2) t is used for a-priori analysis, MetaSim wo conduct more simulations than a post-layout simu tor would be required to. This suggests that incre run time might be an issue of concern. There are th is the frequency-domain Green's function from portints, however. First, in high speed des j to port i at radian frequency-b[a]nd so the avoiding the need for multiple iterations of the inverse Fourier transform Finall  $\chi_{ax} \Delta t = 0.5i/f_1^n$  will save more user time and computer time th

also lead to more efficient use of board space. Seco Based on these techniques, Transimhas bee Heimpi en svery efficient with the human designer' mented [8]. Essentially, the interconnect structures to initial results indicate that the be simulated, are partitioned into a collections for matrion is generally regular and thus a gear, frequency dependent parameter structcharacterization can be obtained with a small num non-linear circuit structures (including of inendome y generated results, as long as the irre

voltages at the terminals of the non-line are at leyment Meta Sim. It is a convolution-based sim and the voltages at the terminals of the lineral hale has solved a number of the problems usual found in such simulators. It has been developed a complete packaging simulator.

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